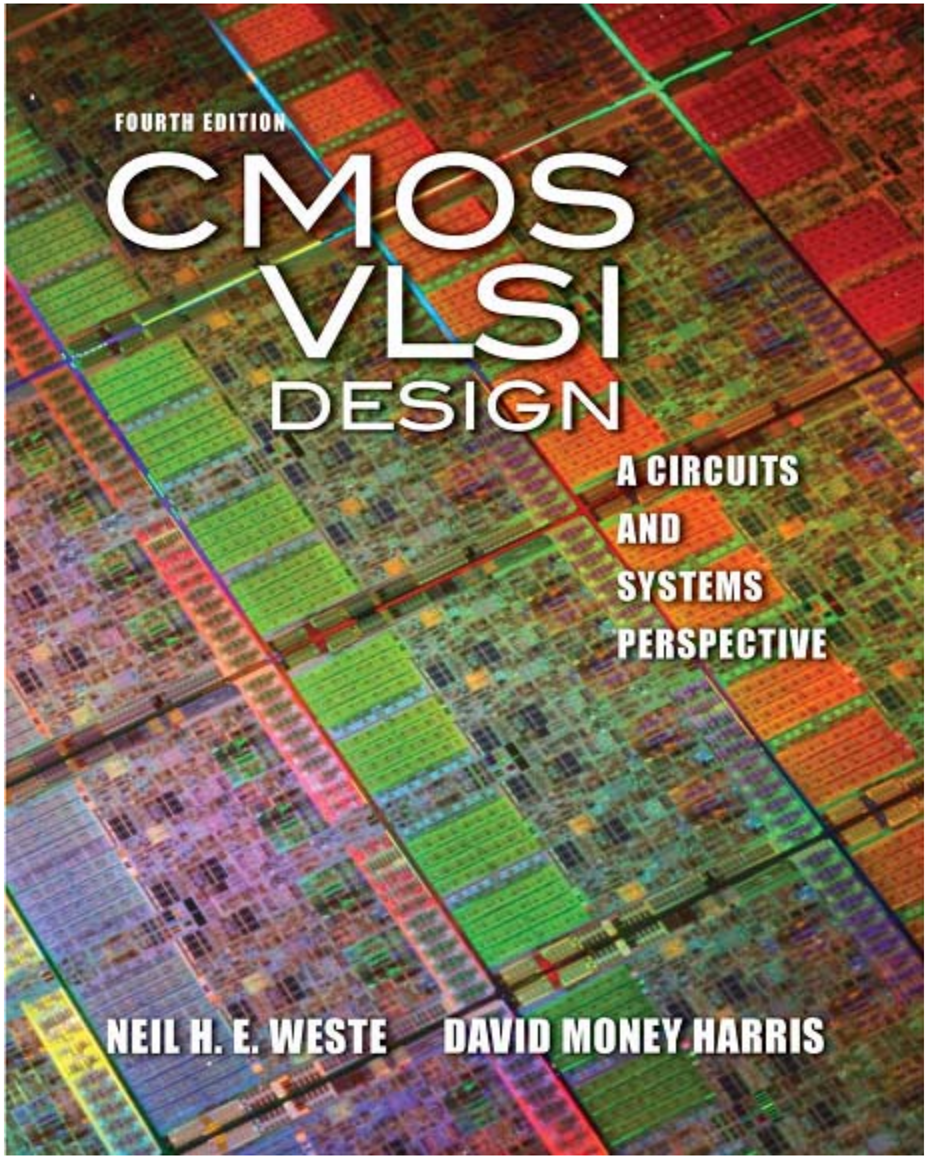


Lecture 14: Wires

Outline

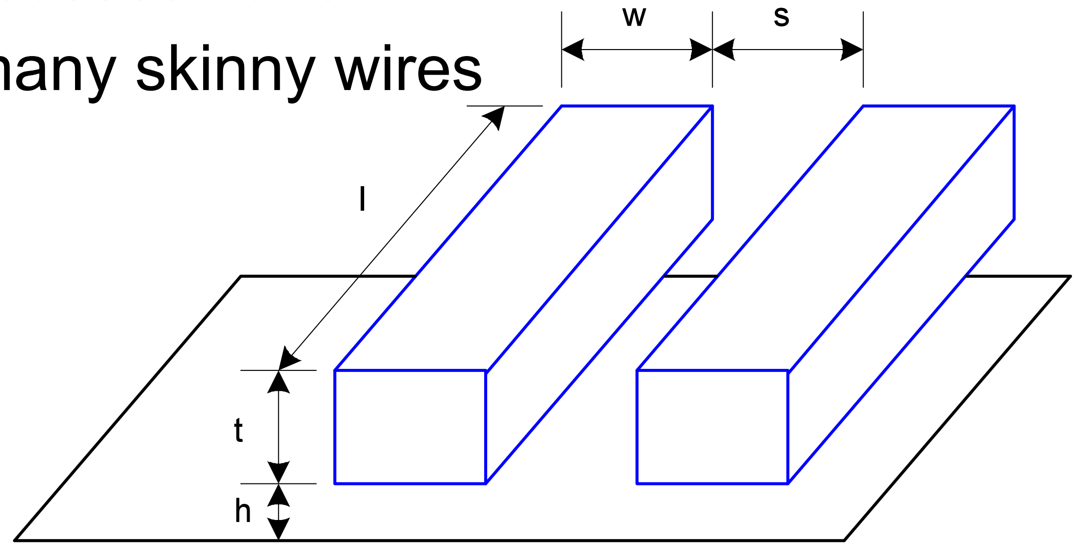
- Introduction
- Interconnect Modeling
  - Wire Resistance
  - Wire Capacitance
- Wire RC Delay
- Crosstalk
- Wire Engineering
- Repeaters

Introduction

- Chips are mostly made of wires called *interconnect*
  - In stick diagram, wires set size
  - Transistors are little things under the wires
  - Many layers of wires
- Wires are as important as transistors
  - Speed
  - Power
  - Noise
- Alternating layers run orthogonally

Wire Geometry

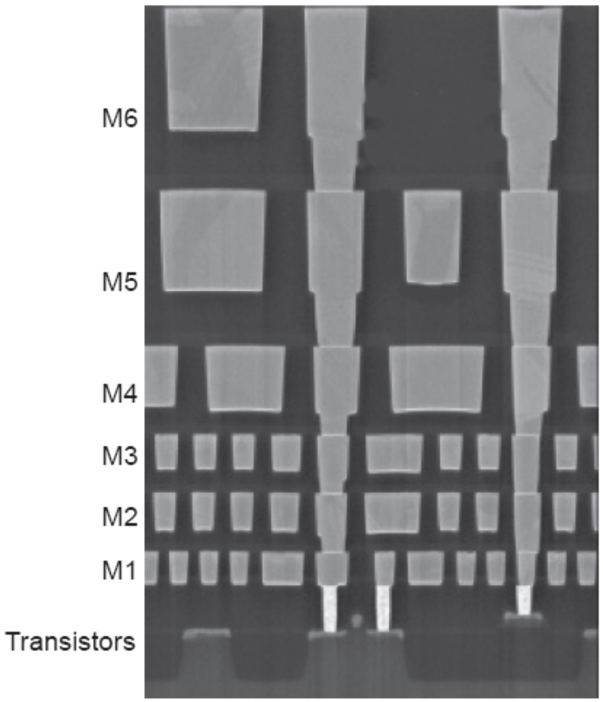
- Pitch = w + s
- Aspect ratio: AR = t/w
  - Old processes had AR << 1
  - Modern processes have AR ≈ 2
    - Pack in many skinny wires

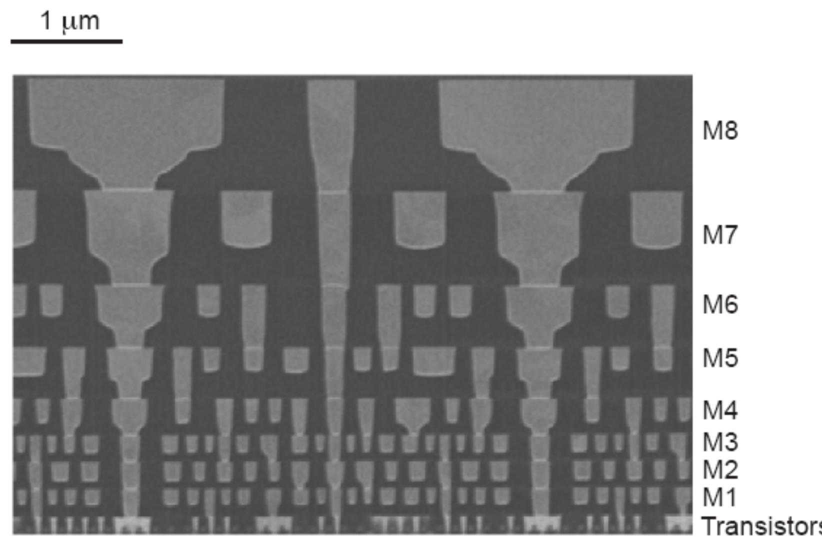


Layer Stack

- AMI 0.6 μm process has 3 metal layers
  - M1 for within-cell routing
  - M2 for vertical routing between cells
  - M3 for horizontal routing between cells
- Modern processes use 6-10+ metal layers
  - M1: thin, narrow (< 3λ)
    - High density cells
  - Mid layers
    - Thicker and wider, (density vs. speed)
  - Top layers: thickest
    - For V_{DD}, GND, clk

Example





Intel 90 nm Stack

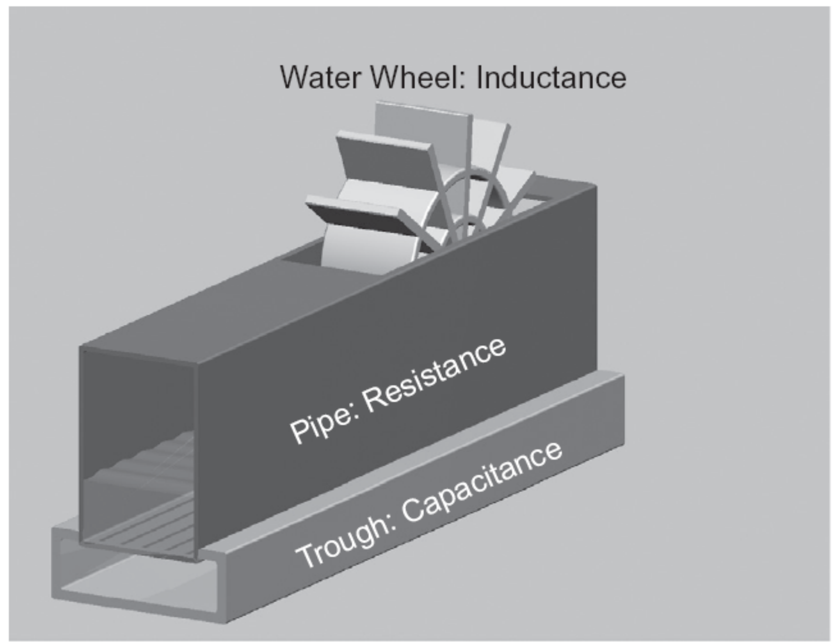
[Thompson02]

Intel 45 nm Stack

[Moon08]

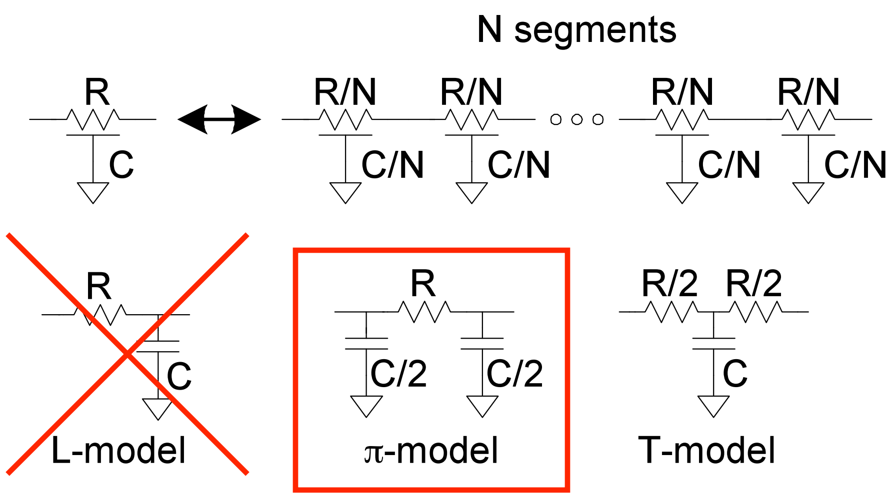
Interconnect Modeling

- Current in a wire is analogous to current in a pipe
  - Resistance: narrow size impedes flow
  - Capacitance: trough under the leaky pipe must fill first
  - Inductance: paddle wheel inertia opposes changes in flow rate
    - Negligible for most wires



Lumped Element Models

- Wires are a distributed system
  - Approximate with lumped element models



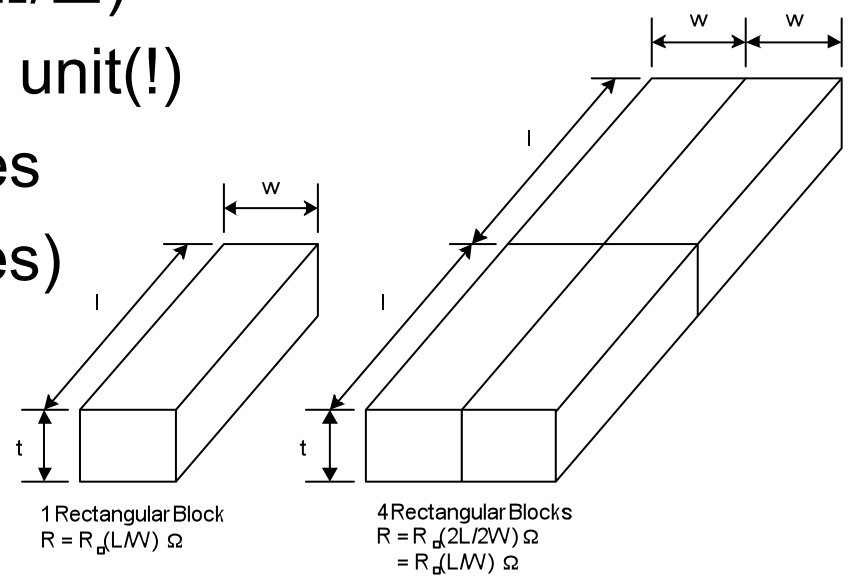
- 3-segment π-model is accurate to 3% in simulation
- L-model needs 100 segments for same accuracy!
- Use single segment π-model for Elmore delay

Wire Resistance

- ρ = *resistivity* (Ω*m)

  $R =$

- $R_{\square}$ = *sheet resistance* ($\Omega/\square$)
  - $\square$ is a dimensionless unit(!)
- Count number of squares
  - $R = R_{\square}$ * (# of squares)



Choice of Metals

- Until 180 nm generation, most wires were aluminum
- Contemporary processes normally use copper
  - Cu atoms diffuse into silicon and damage FETs
  - Must be surrounded by a diffusion barrier

Metal	**Bulk resistivity ($\mu\Omega \cdot$ cm)**
Silver (Ag)	1.6
Copper (Cu)	1.7
Gold (Au)	2.2
Aluminum (Al)	2.8
Tungsten (W)	5.3
Titanium (Ti)	43.0

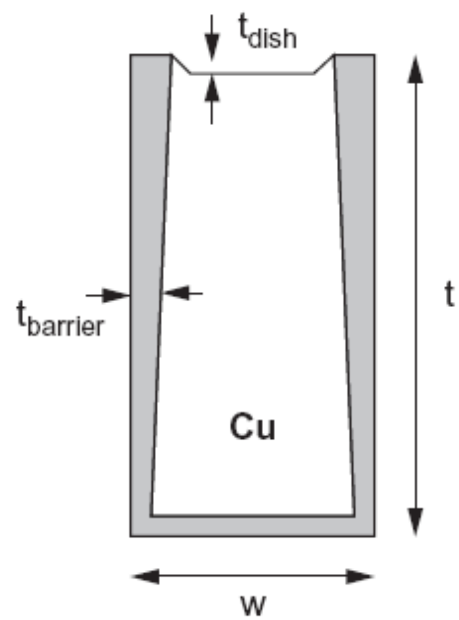
Contacts Resistance

- Contacts and vias also have 2-20 Ω
- Use many contacts for lower R
  - Many small contacts for current crowding around periphery

Copper Issues

- Copper wires diffusion barrier has high resistance
- Copper is also prone to *dishing* during polishing
- Effective resistance is higher

$$R = \frac{\rho}{\left(t - t_{\text{dish}} - t_{\text{barrier}}\right)} \frac{l}{\left(w - 2t_{\text{barrier}}\right)}$$



Example

- Compute the sheet resistance of a 0.22 μm thick Cu wire in a 65 nm process. Ignore dishing.

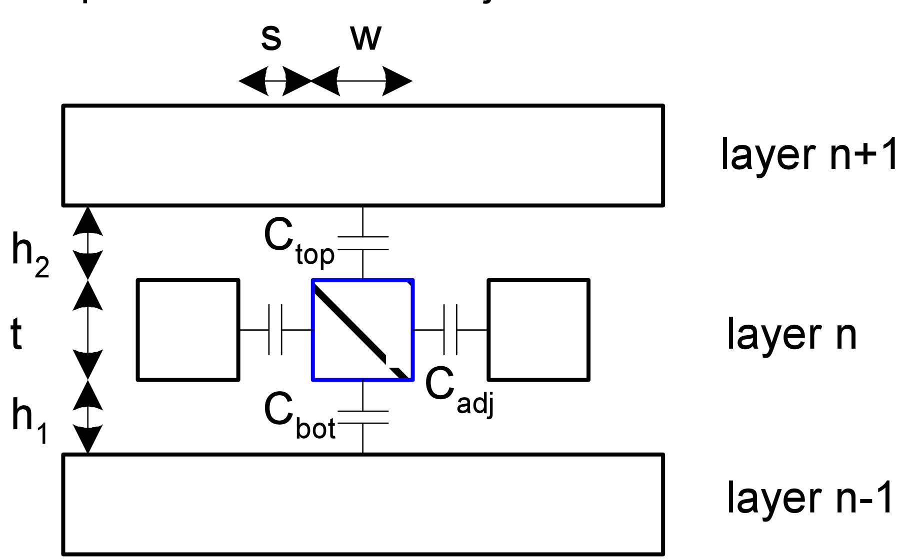
$$R_{\square} =$$

- Find the total resistance if the wire is 0.125 μm wide and 1 mm long. Ignore the barrier layer.

$$R =$$

Wire Capacitance

- Wire has capacitance per unit length
  - To neighbors
  - To layers above and below
- $C_{total} = C_{top} + C_{bot} + 2C_{adj}$



Capacitance Trends

- Parallel plate equation: $C = \varepsilon_{ox} A/d$
  - Wires are not parallel plates, but obey trends
  - Increasing area (W, t) increases capacitance
  - Increasing distance (s, h) decreases capacitance
- Dielectric constant
  - $\varepsilon_{ox} = k\varepsilon_0$
    - $\varepsilon_0 = 8.85 \times 10^{-14}$ F/cm
    - $k = 3.9$ for SiO_2
- Processes are starting to use low-k dielectrics
  - $k \approx 3$ (or less) as dielectrics use air pockets

Capacitance Formula

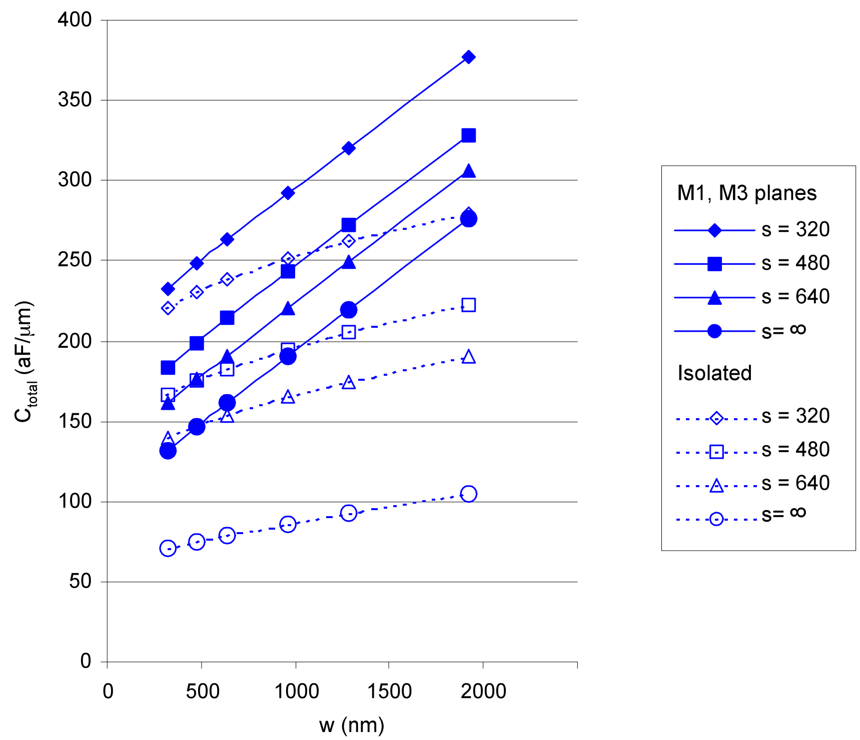
- Capacitance of a line without neighbors can be approximated as

$$C_{tot} = \varepsilon_{\mathrm{ox}} l \left[\frac{w}{h} + 0.77 + 1.06 \left(\frac{w}{h} \right)^{0.25} + 1.06 \left(\frac{t}{h} \right)^{0.5} \right]$$

- This empirical formula is accurate to 6% for AR < 3.3

M2 Capacitance Data

- Typical dense wires have ~ 0.2 fF/μm
  - Compare to 1-2 fF/μm for gate capacitance



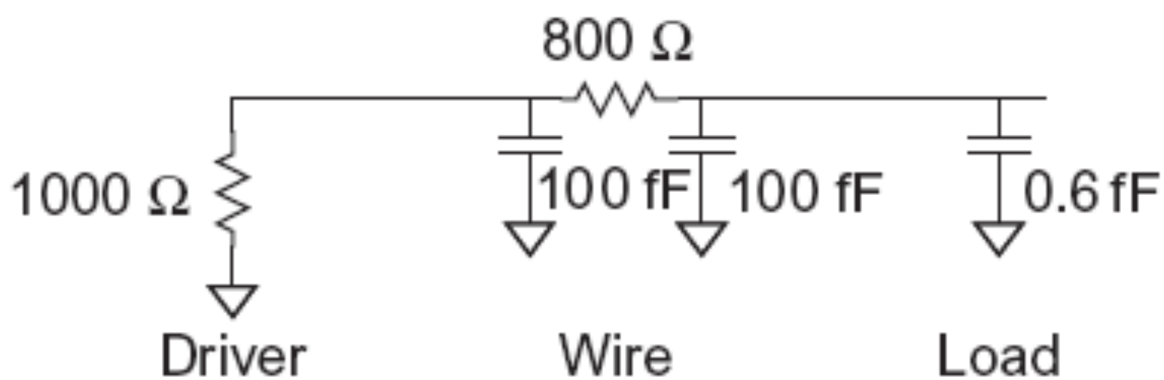
Diffusion & Polysilicon

- Diffusion capacitance is very high (1-2 fF/μm)
  - Comparable to gate capacitance
  - Diffusion also has high resistance
  - Avoid using diffusion *runners* for wires!
- Polysilicon has lower C but high R
  - Use for transistor gates
  - Occasionally for very short wires between gates

Wire RC Delay

- Estimate the delay of a 10x inverter driving a 2x inverter at the end of the 1 mm wire. Assume wire capacitance is 0.2 fF/μm and that a unit-sized inverter has R = 10 KΩ and C = 0.1 fF.

  – t_{pd} =



Wire Energy

- Estimate the energy per unit length to send a bit of information (one rising and one falling transition) in a CMOS process.

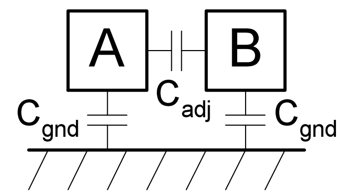
- E =

Crosstalk

- A capacitor does not like to change its voltage instantaneously.
- A wire has high capacitance to its neighbor.
  - When the neighbor switches from 1-> 0 or 0->1, the wire tends to switch too.
  - Called capacitive *coupling* or *crosstalk*.
- Crosstalk effects
  - Noise on nonswitching wires
  - Increased delay on switching wires

Crosstalk Delay

- Assume layers above and below on average are quiet
  - Second terminal of capacitor can be ignored
  - Model as $C_{gnd} = C_{top} + C_{bot}$
- Effective C_{adj} depends on behavior of neighbors
  - *Miller effect*

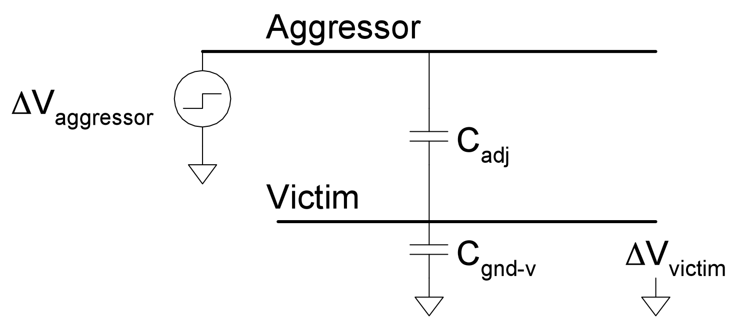


B	**ΔV**	**$C_{eff(A)}$**	**MCF**
Constant			
Switching with A			
Switching opposite A			

Crosstalk Noise

- Crosstalk causes noise on nonswitching wires
- If victim is floating:
  - model as capacitive voltage divider

$$\Delta V_{victim} = \frac{C_{adj}}{C_{gnd-v} + C_{adj}} \Delta V_{aggressor}$$

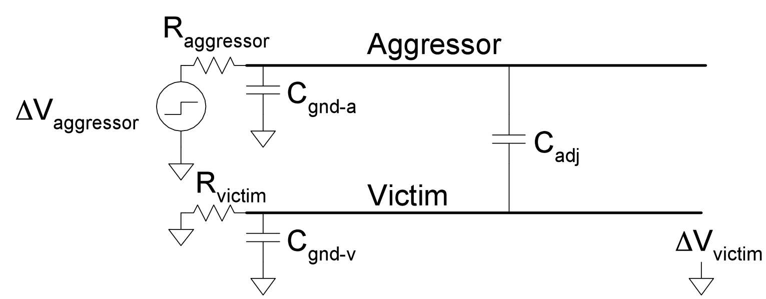


Driven Victims

- Usually victim is driven by a gate that fights noise
  - Noise depends on relative resistances
  - Victim driver is in linear region, agg. in saturation
  - If sizes are same, $R_{aggressor}$ = 2-4 x R_{victim}

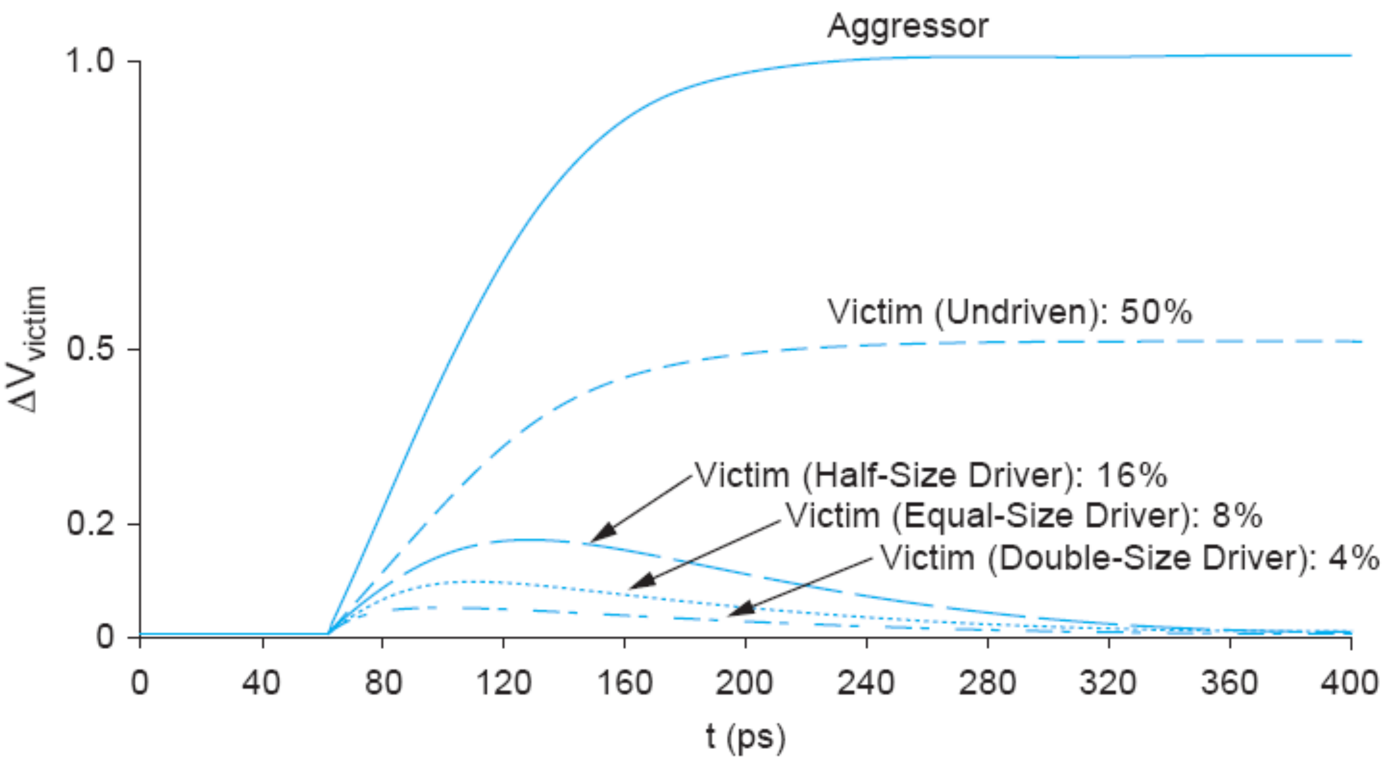
$$\Delta V_{victim} = \frac{C_{adj}}{C_{gnd-v} + C_{adj}} \frac{1}{1+k} \Delta V_{aggressor}$$

$$k = \frac{\tau_{aggressor}}{\tau_{victim}} = \frac{R_{aggressor}\left(C_{gnd-a} + C_{adj}\right)}{R_{victim}\left(C_{gnd-v} + C_{adj}\right)}$$



Coupling Waveforms

- Simulated coupling for $C_{adj} = C_{victim}$

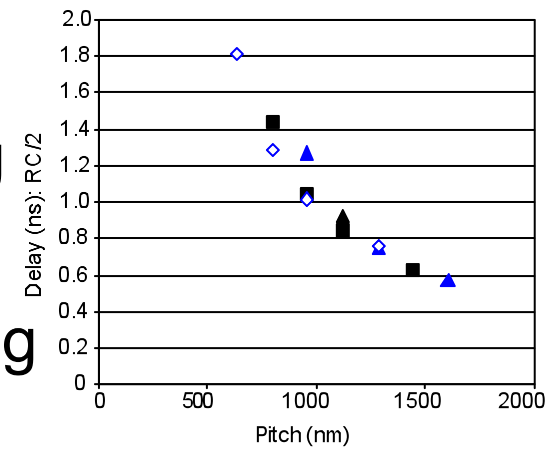


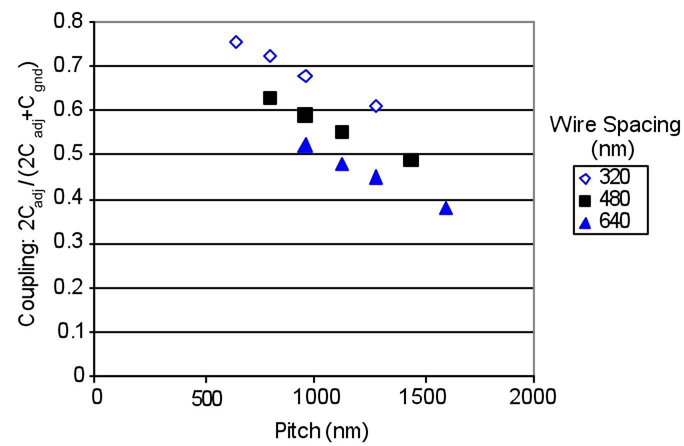
Noise Implications

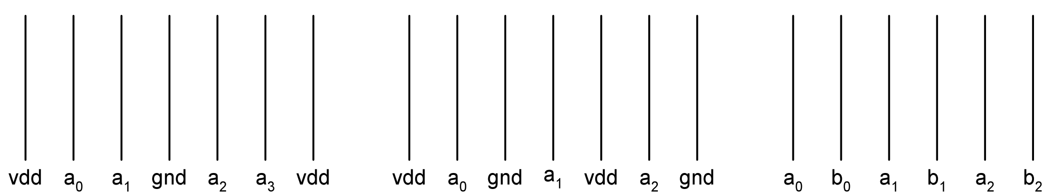
- *So what* if we have noise?
- If the noise is less than the noise margin, nothing happens
- Static CMOS logic will eventually settle to correct output even if disturbed by large noise spikes
  - But glitches cause extra delay
  - Also cause extra power from false transitions
- Dynamic logic never recovers from glitches
- Memories and other sensitive circuits also can produce the wrong answer

Wire Engineering

- Goal: achieve delay, area, power goals with acceptable noise
- Degrees of freedom:
  - Width
  - Spacing
  - Layer
  - Shielding

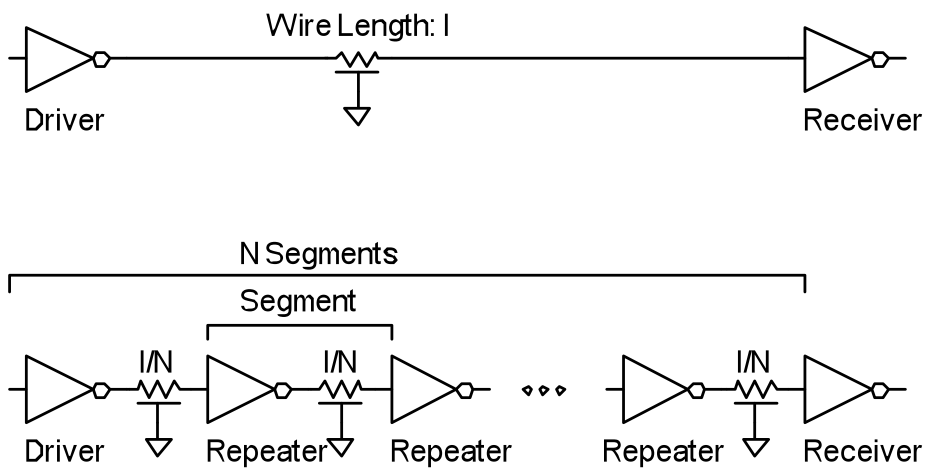






Repeaters

- R and C are proportional to
- RC delay is proportional to
  - Unacceptably great for long wires
- Break long wires into N shorter segments
  - Drive each one with an inverter or buffer



Repeater Design

- How many repeaters should we use?
- How large should each one be?
- Equivalent Circuit
  - Wire length *l/N*
    - Wire Capacitance C_w**l/N*, Resistance R_w**l*/N
  - Inverter width W (nMOS = W, pMOS = 2W)
    - Gate Capacitance C'*W, Resistance R/W

Repeater Results

- Write equation for Elmore Delay
  - Differentiate with respect to W and N
  - Set equal to 0, solve

$$\frac{l}{N} = \sqrt{\frac{2RC'}{R_w C_w}}$$

$$\frac{t_{pd}}{l} = \left(2 + \sqrt{2}\right)\sqrt{RC'R_w C_w}$$

~40 ps/mm

in 65 nm process

$$W = \sqrt{\frac{RC_w}{R_w C'}}$$

Repeater Energy

- Energy / length ≈ $1.87C_wV_{DD}^2$
  - 87% premium over unrepeated wires
  - The extra power is consumed in the large repeaters
- If the repeaters are downsized for minimum EDP:
  - Energy premium is only 30%
  - Delay increases by 14% from min delay